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Measurements

If you have a smart project, you can say "I'm an engineer"

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Lecture 3

Staff boarder

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Measurements

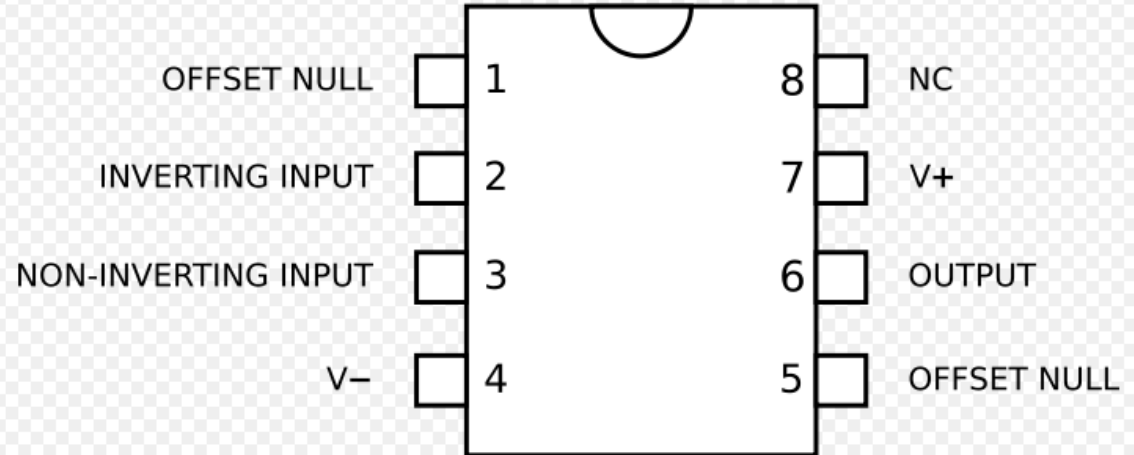
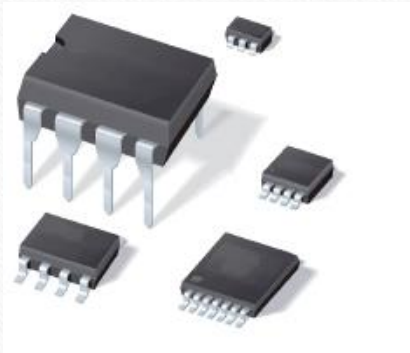
- **Lecture aims:**
 - Develop the Signal Conditioning for measurements
 - Formulate advanced problems for Filters

Signal Conditioning

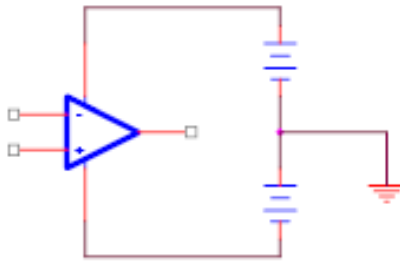
- The signal output from the sensor is usually small (in mV) and noisy
- Signal Conditioning is the process of filtering and amplifying this signal in order to process it to the computer
- Amplification was studied and therefore will not be covered here
- In this section, we will study analog filters

Amplifiers

Operational amplifier

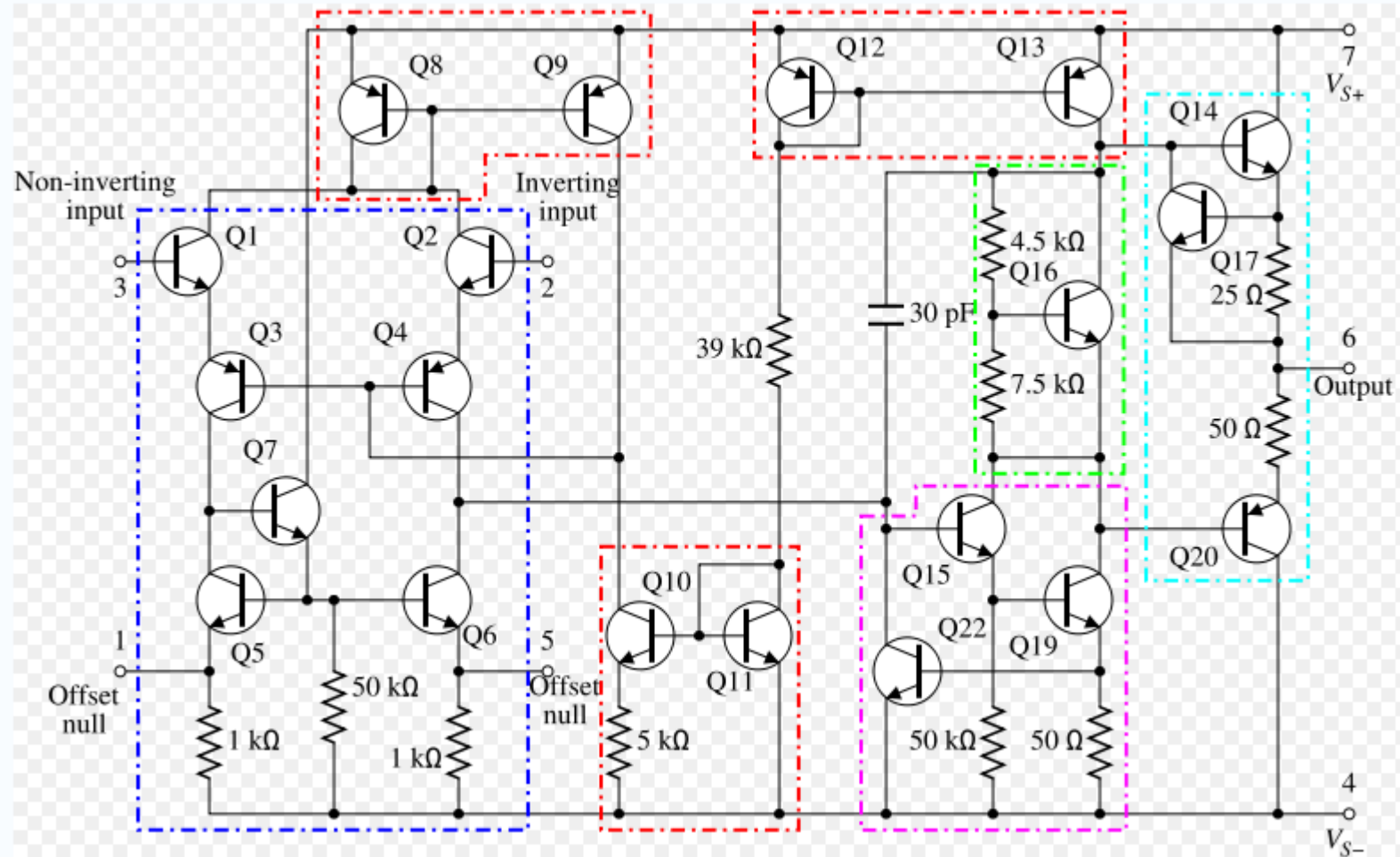


Power Connections



Amplifiers

Image: OpAmp Transistor Level



Amplifiers

Operational amplifier

$$\frac{(E-) + E_1}{R_1} = \frac{E_o - ((E-) + E_1)}{R_o}$$

but:

$$E_o = -A \cdot (E-)$$

$$(E-) = \frac{-E_o}{A}$$

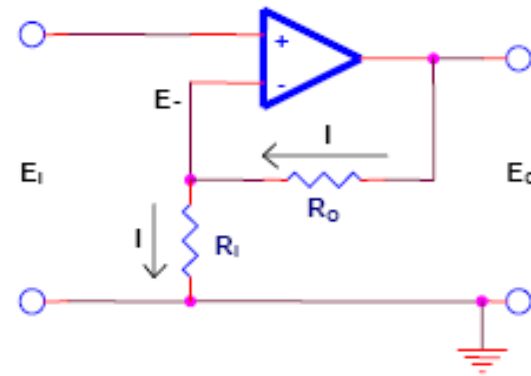
$$\frac{E_1}{R_1} = \frac{E_o - E_1}{R_o}$$

Solving:

$$E_1(R_o + R_1) = E_o R_1$$

$$\frac{E_o}{E_1} = \frac{R_o + R_1}{R_1} = 1 + \frac{R_o}{R_1}$$

THE NON-INVERTING AMPLIFIER



Amplifiers

Let the voltage at the inverting input be E_- and the open loop operational amplifier gain be A (ideally, $A = \infty$).

Since equal currents flow in R_o and R_i :

$$\frac{E_i - (E_-)}{R_i} + \frac{E_o - (E_-)}{R_o} = 0$$

But by definition:

$$E_o = A \cdot (E_-)$$

$$(E_-) = -\frac{E_o}{A}$$

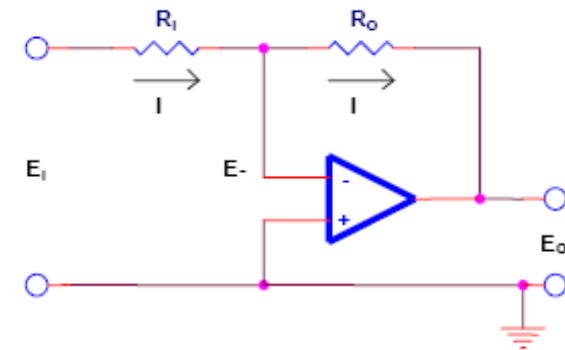
Letting A go to infinity, E_- approaches zero and:

$$\frac{E_i}{R_i} + \frac{E_o}{R_o} = 0$$

or:

$$\frac{E_o}{E_i} = -\frac{R_o}{R_i}$$

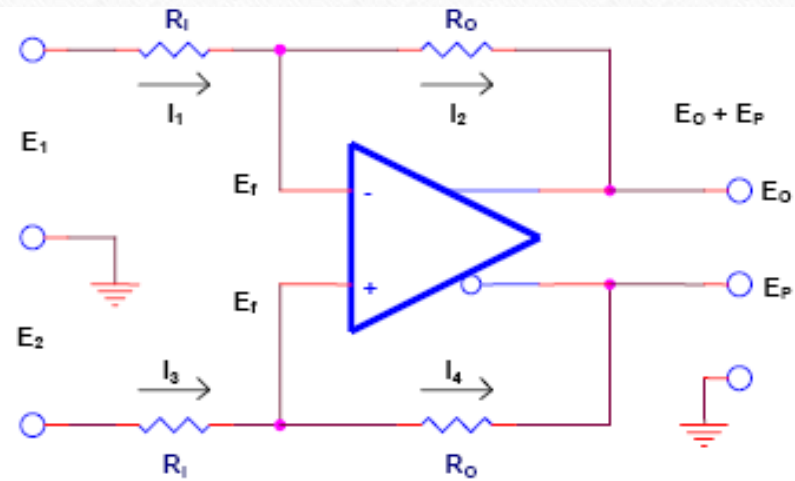
THE INVERTING AMPLIFIER



Amplifiers

THE DIFFERENTIAL (BALANCED) OUTPUT AMPLIFIER

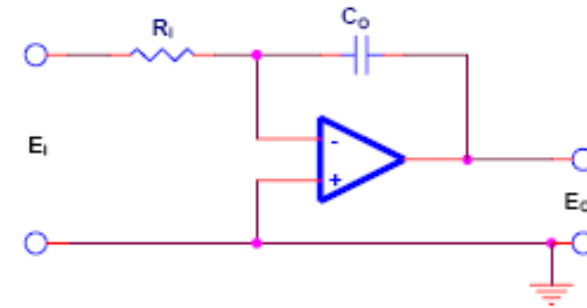
$$E_o = \frac{R_o}{R_i}(E_2 - E_1)$$



Amplifiers

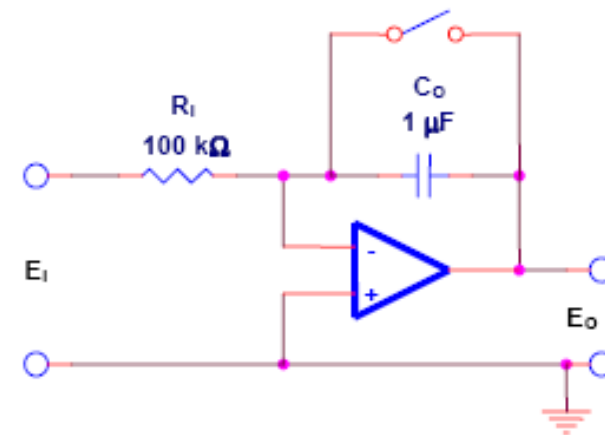
INTEGRATORS

$$E_o = \frac{-Z_o}{Z_i} E_i = \frac{-E_i}{R_i C_{op}} = -\frac{1}{R_i C_o} E_i dt$$



PRACTICAL INTEGRATORS

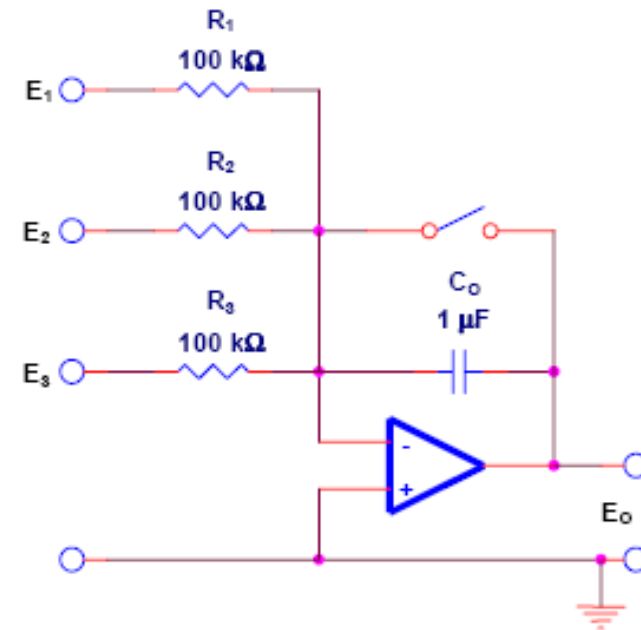
Simple integrator circuits operate successfully but current offset is stored in the feedback capacitor causing output voltage error. This is corrected with low drift chopper stabilized amplifier and/or current biasing networks.



Amplifiers

Summing Integrator

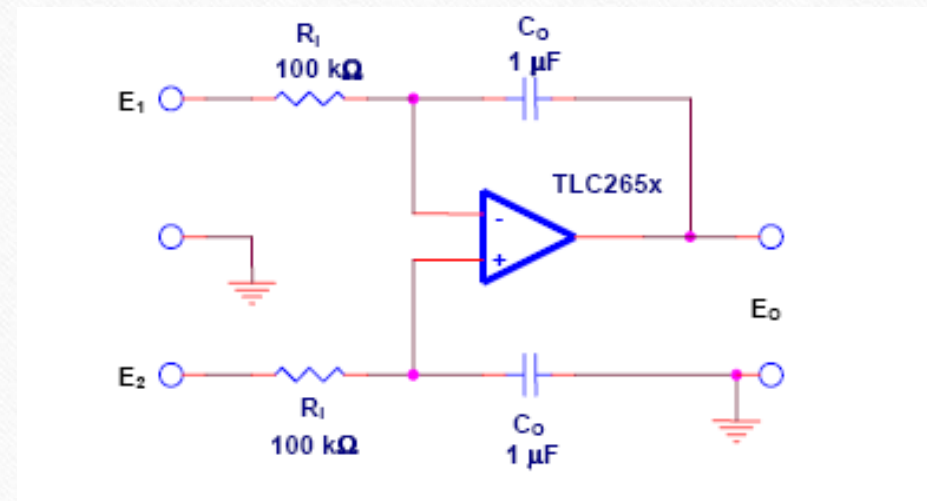
$$E_o = \frac{-1}{RC} (E_1 + E_2 + E_3)dt$$
$$= -10 (E_1 + E_2 + E_3)dt$$



Amplifiers

Differential Integrator

$$E_o = \frac{-1}{R_1 C_o} (E_1 - E_2) dt$$
$$= 10 (E_2 - E_1) dt$$

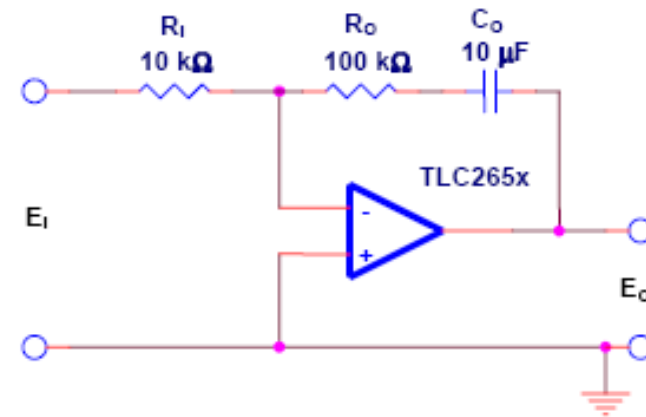


Amplifiers

Augmenting Integrator

Sums the input signal and its time integral.

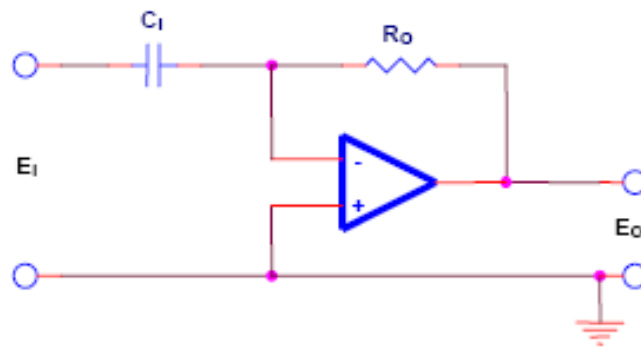
$$E_o = \frac{-R_o E_i}{R_i} - \frac{1}{C_o R_i} \int E_i dt$$
$$= -10E_i - \int E_i dt$$



Amplifiers

DIFFERENTIATORS

For the differentiator; $Z_o = R_o$, $Z_i = \frac{1}{C_p}$



$$E_o = \frac{-Z_o}{Z_i} E_i = -R_o C_i \frac{dE_i}{dt}$$

Amplifiers

THE VOLTAGE SUMMER

Assume the ideal summing point restraints:

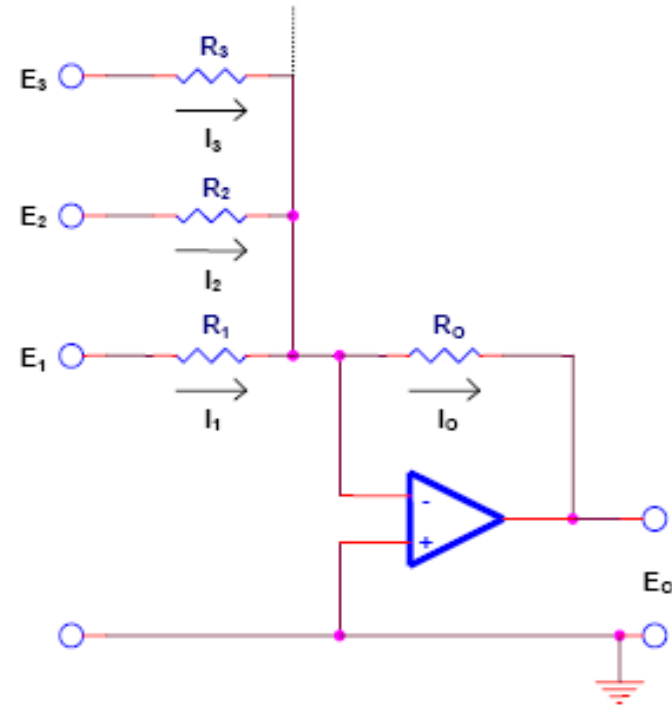
1. Pin (1) is at ground potential
2. $-I_o + I_1 + I_2 + I_3 + \dots = 0$

The current, I_o , is given by:

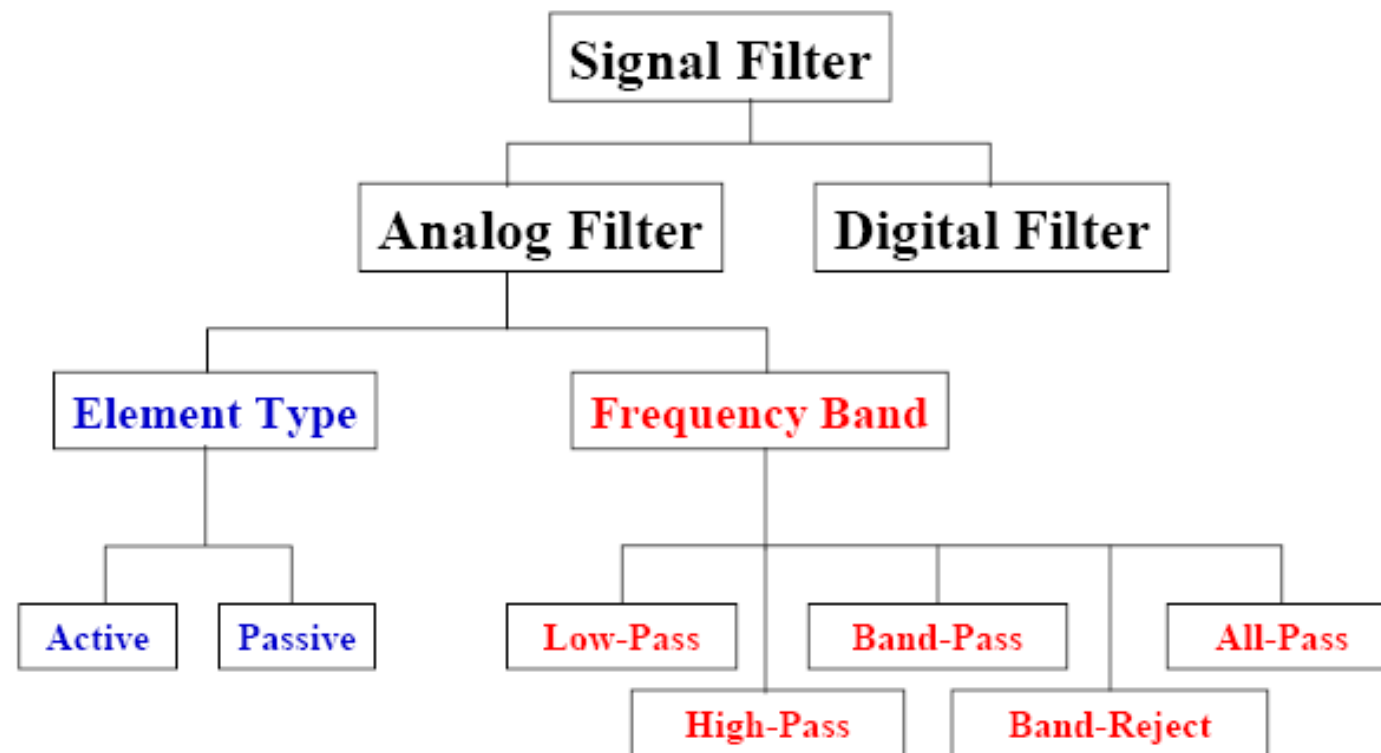
$$I_o = \frac{-E_o}{R_o} = \frac{E_1}{R_1} + \frac{E_2}{R_2} + \frac{E_3}{R_3} + \dots$$

So that:

$$E_o = -R_o \left(\frac{E_1}{R_1} + \frac{E_2}{R_2} + \frac{E_3}{R_3} + \dots \right)$$



Classification of Filters

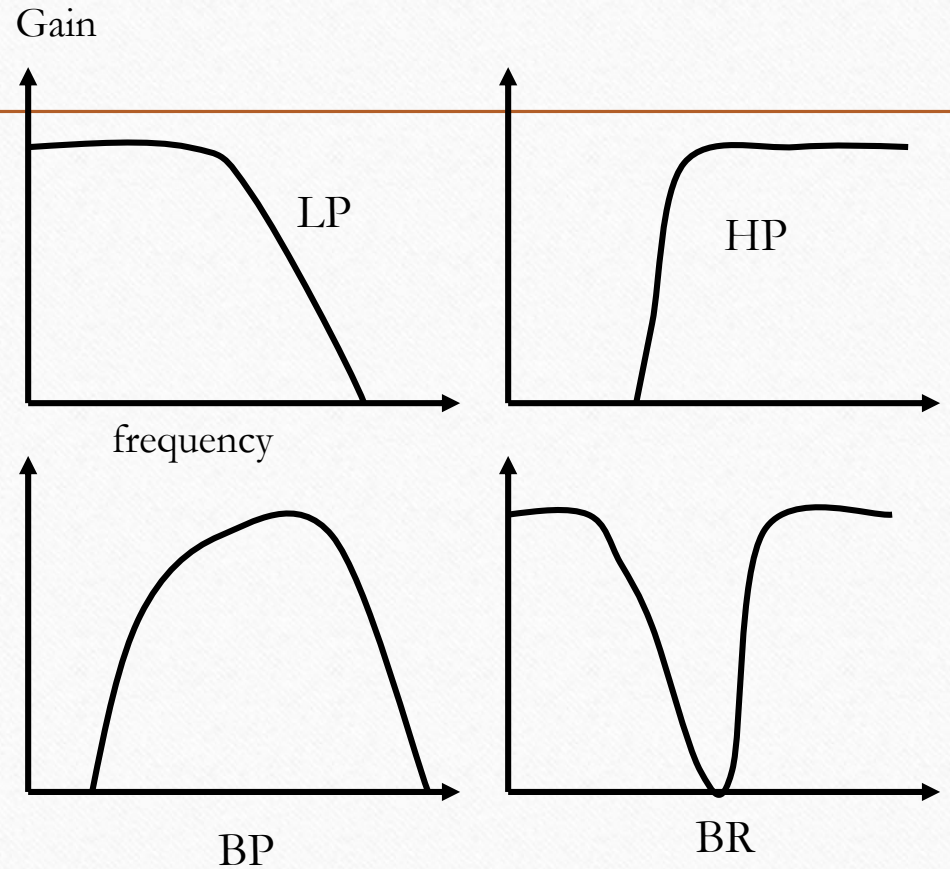


Analog Filters

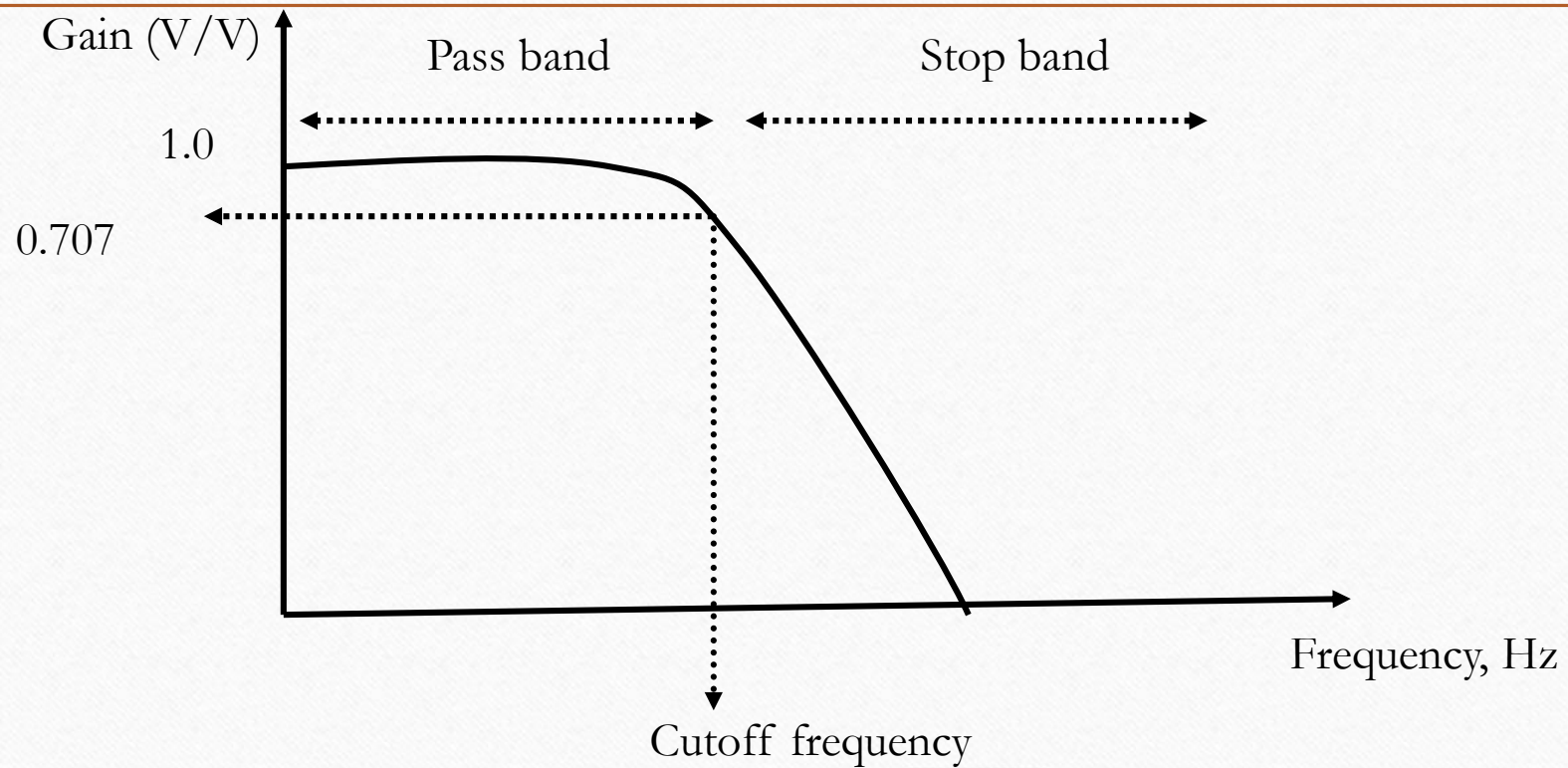
- Electrical filters are designed to eliminate unwanted frequencies (i.e. noise)
- Filters are divided into Passive and Active filters.
- Passive filters use only resistors, capacitors, and inductors
- Active filters contain op-amps and therefore can amplify the signal

Types of filters

- Filters can be divided into 4 types:
 - Low Pass (LP) filters
 - High Pass (HP) filters
 - Band Pass (BP) filters
 - Notch or Band Reject (BR) filters



Low Pass Filter



Filter Characteristics

- Ideally, Low Pass Filter passes frequencies $<$ cutoff frequency (f_c) and blocks frequencies $> f_c$
- **Cutoff Frequency** is the frequency at which the voltage gain of the filter drops to 0.707 of its maximum value
- The range of frequencies that pass the filter are called **Passband**
- The range of frequencies that do Not pass through the filter are called **Stopband**
- For passive filters, the gain in the passband is equal to 1

Passive LP filter example: RC Circuit

The transfer function can be found using voltage divider rule

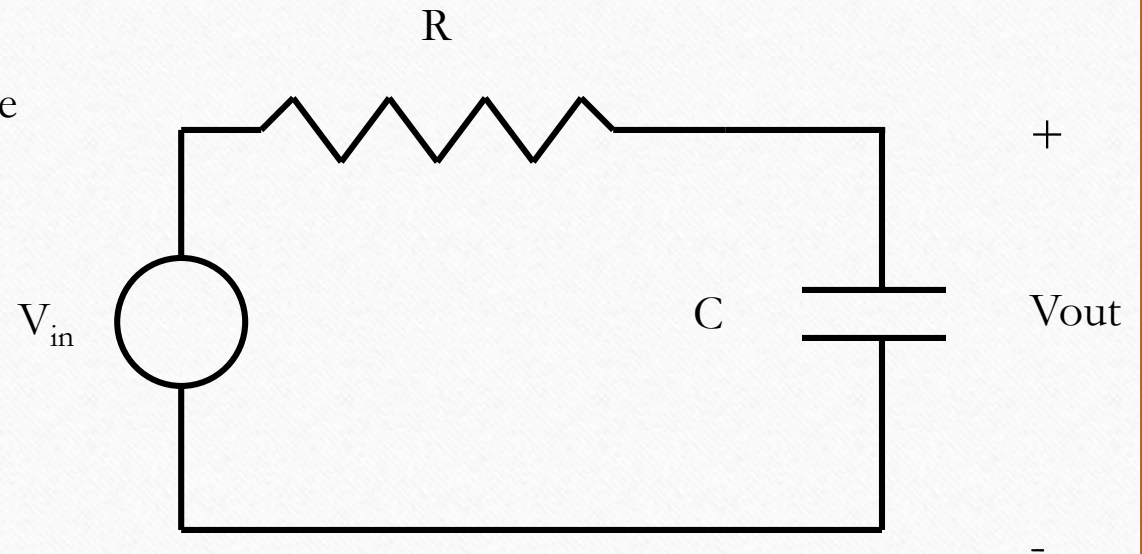
$$H(j\omega) = \frac{1/j\omega C}{1/j\omega C + R}$$

The magnitude is found using

$$|H(j\omega)| = \sqrt{\operatorname{Re}\{H(j\omega)\}^2 + \operatorname{Im}\{H(j\omega)\}^2}$$

The phase is found using

$$\Phi(j\omega) = \tan^{-1} \frac{\operatorname{Im}\{H(j\omega)\}}{\operatorname{Re}\{H(j\omega)\}}$$



RC circuit example

This in turn gives the magnitude to be equal to

$$|H(j\omega)| = \frac{1/RC}{\sqrt{\omega^2 + (1/RC)^2}}$$

In order to get, f_c , we solve the following equation

$$|H(j\omega_c)| = \frac{1}{\sqrt{2}} H_{\max} = \frac{1}{\sqrt{2}} \quad (1)$$

$$\Rightarrow \frac{1}{\sqrt{2}} = \frac{1/RC}{\sqrt{\omega_c^2 + (1/RC)^2}}$$

$$\Rightarrow \omega_c = 1/RC$$

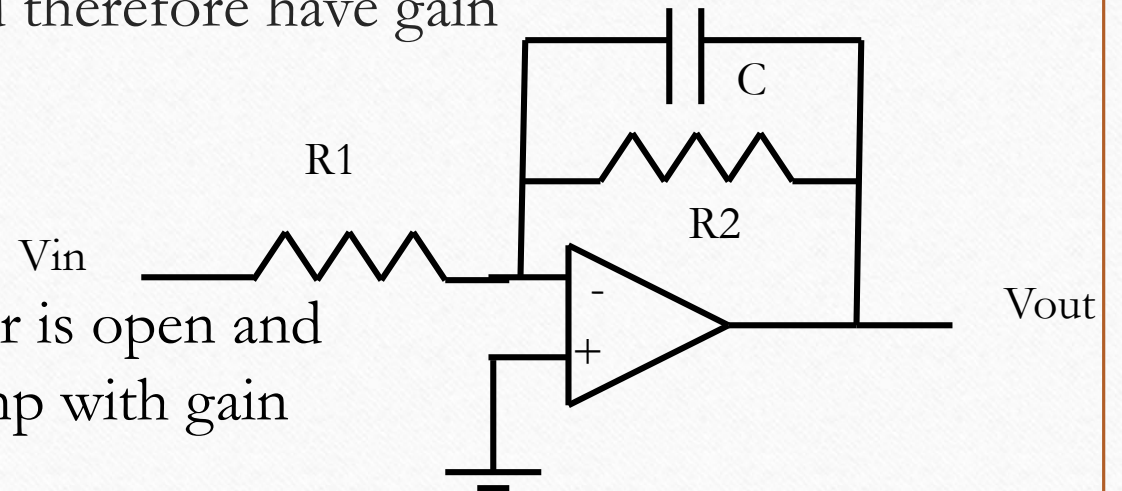
$$\Rightarrow f_c = 1/2\pi RC$$

Active Filters: LP

- Active filters use op-amps in their design and therefore have gain
- An active low pass filter is shown below

When input frequency is 0, then the capacitor is open and the circuit becomes a regular inverting op-amp with gain $R2/R1$

When input frequency is infinite, the capacitor is short and the Output voltage becomes zero



Active Filters: LP

The transfer function, magnitude, and the cutoff frequency of the circuit shown is derived to be

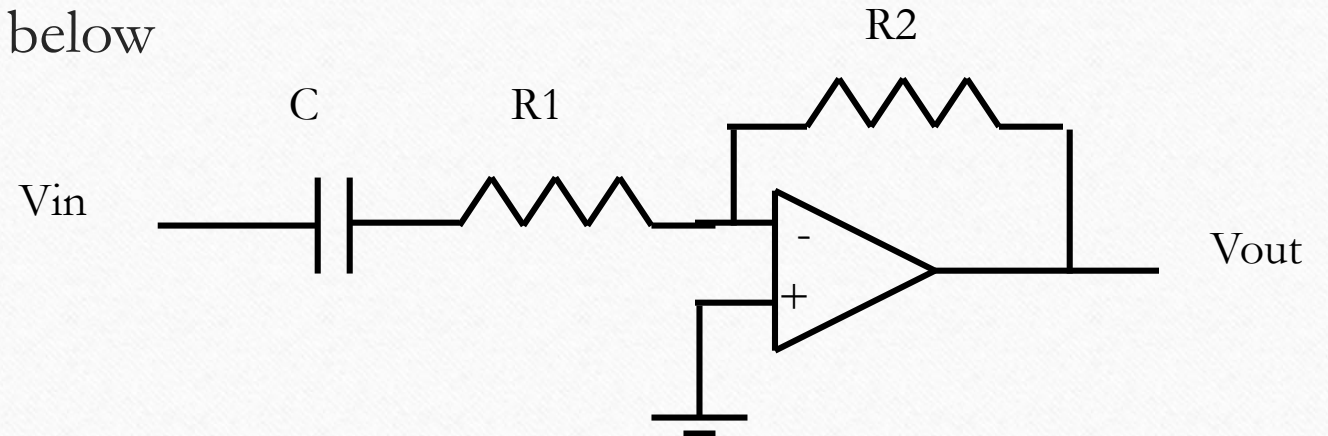
$$H(j\omega) = \frac{-Z_{out}}{Z_{in}} = \frac{-R_2 \parallel (1/j\omega C)}{R_1}$$

$$|H(j\omega)| = \frac{R_2}{R_1} \frac{1}{\sqrt{(R_2 C \omega)^2 + 1}}$$

$$\omega_c = 1/R_2 C$$

Active Filters: HP

An active high pass filter is shown below



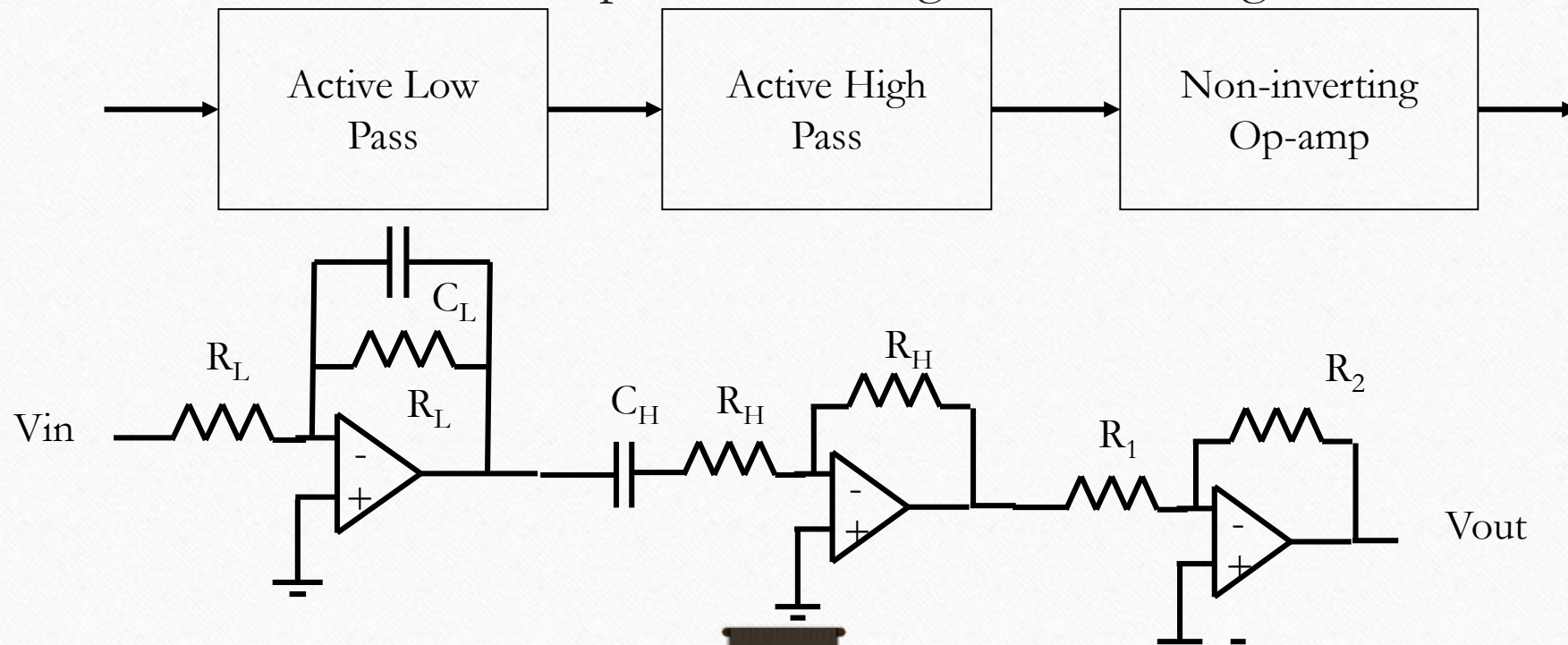
The transfer function and the cutoff frequency of the circuit shown is derived to be

$$H(j\omega) = \frac{-R_2}{R_1} \frac{j\omega}{j\omega + 1/R_1C}$$

$$\omega_c = 1/R_1C$$

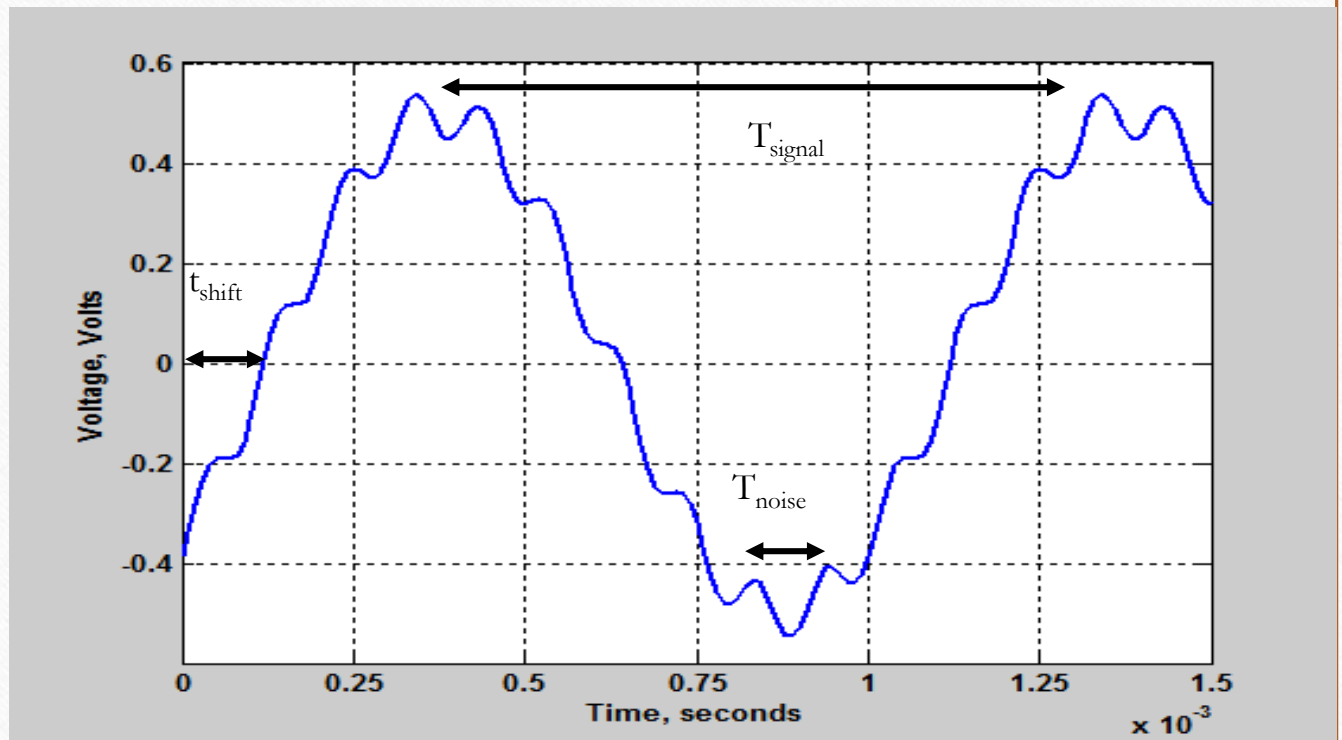
Active Filters: Band Pass

We can construct an active band pass filter using 3 cascade stages



Active Filter Design Example

Problem Statement: A transducer is used as an input to a PC. The measured signal is sinusoidal with high frequency noise added. The signal is shown in figure below. Design an analog signal conditioning circuit to filter out the noise and give a gain of 10 for the sinusoidal signal



Example continued

The original signal had a period, T_{signal} , of 1 ms $\Rightarrow F_{\text{signal}} = 1,000$ Hz

The time shift, t_{shift} , is equal to 0.125 ms \Rightarrow phase shift is $0.125\text{ms} \times 360^\circ / 1\text{ms}$
 $= 45$ degrees

Then, the signal can be represented as

$$x(t) = 0.5 \sin(2\pi(1,000)t - 45) + \text{noise}$$

Amplitude

frequency

Phase shift

The noise frequency is calculated to be around 8 KHz

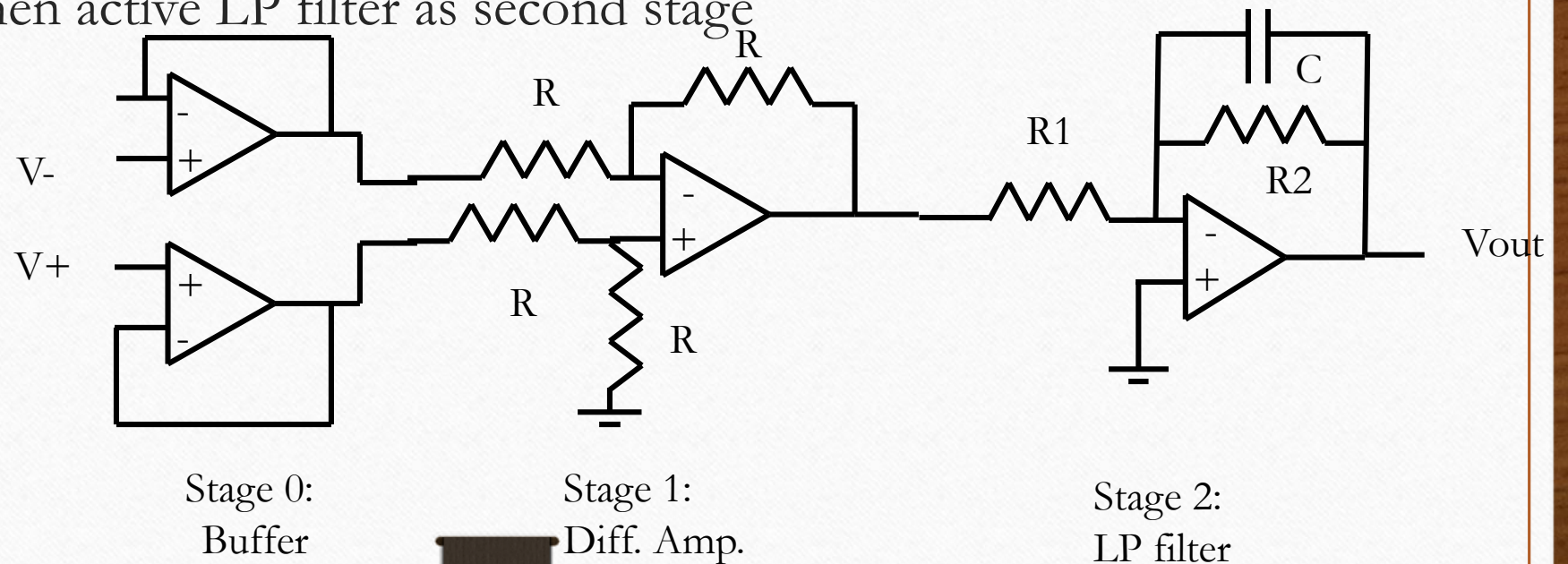
And since the signal frequency is 1 KHz

\Rightarrow need LP filter with cutoff frequency around 4 KHz

Example continued

There are many ways to design this filter. Here is one way:

Since the transducer signal is differential => use a difference amplifier at first stage and then active LP filter as second stage



Example continued

Use $R = 1\text{K}\Omega$ at the buffers (no gain)

The cutoff frequency is $\omega_c = 1/CR_2$

Let $C = 0.1\mu\text{F}$ and since $\omega_c = 2,000\pi \Rightarrow R_2 = 398\Omega$

Gain is equal $R_2/R_1 = 10 \Rightarrow R_1 = 39.8\Omega$

Summary

- Signal Conditioning is the process of filtering and amplification of signals
- Filters are divided into 4 types: low pass, high pass, band pass, and band reject
- Cutoff frequency, pass band and stop band are important filter characteristics
- Active filters use op-amps in their design and therefore create impedance isolation between stages. They have better accuracy than passive filters